# Mobile Robot Navigation – some issues in controller design and implementation

# L. Huang

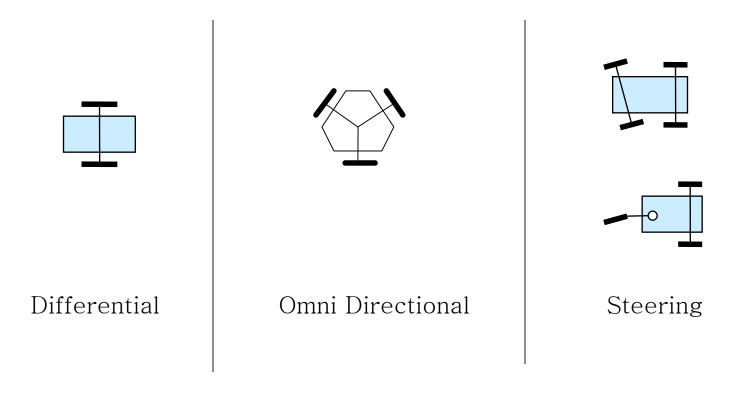
# School of Engineering and Advanced Technology Massey University

# Outlines

- 1. Introduction
- 2. Target tracking control schemes based on
  - Lyapunov method
  - Potential field method
- 3. Speed control
- 4. Conclusion

# Introduction

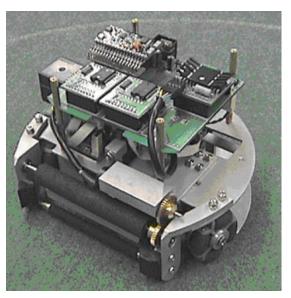
A wheeled mobile robot (WMR) can be driven by wheels in various formations:

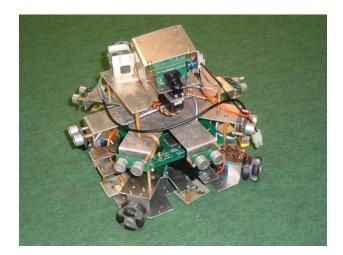






#### Differential Wheel Robot

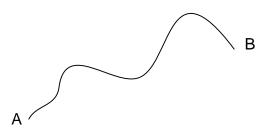




Omni Wheel Robot

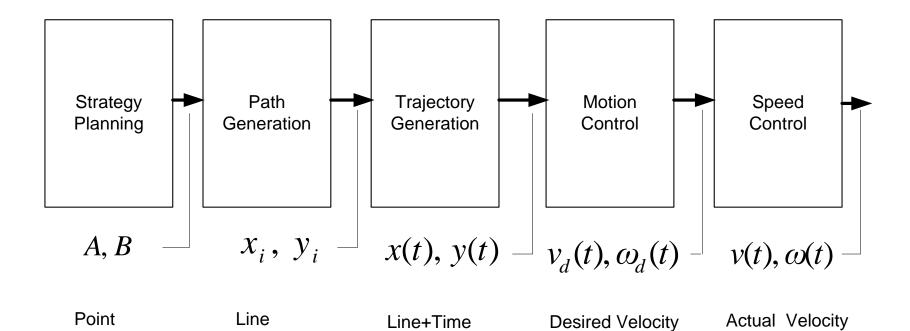
Two basic issues:

1. How to move a robot from posture A to posture B stand alone ?



2. How to determine postures A and B for a robot when a group of robots performing a task (such as soccer playing) ?





## Differential wheel driven robot (no-holonomic):

➢Robot's posture (Cartesian coordinates) cannot be stablized by time-invariant feedback control or smooth state feedback control (*Brockett R. W.* etc.).

Stabilization problem was solved by discontinuous or time varying control in Cartesian space (*Campion G. B., Samson C.* etc.)

➢Asymptotic stabilization through smooth state feedback was achieved by Lyapunov design in Polar coordinates – the system is singular in origin, thus avoids the Brockett's condition (*Aicardi M.* etc ).

➤Trajectory tracking control is easier to achieve and is more significant in practice (desired velocity nonzero) (*Caudaus De Wit*, *De Luca A* etc.).

### Omni-wheel driven robot

 It is fully linearisable for the controller design (D'Andrea-Novel etc.)
 Dynamic optimal control was implemented (Kalmar-

Nagy etc.)

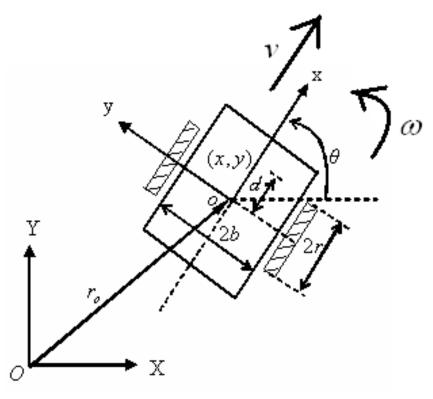
## Robot modeled as a point-mass

Potential field method was used for robot path planning (*Y.Koren and J. Borenstein*)

### Issues to be addressed

Application of Lyapunov-based and potential field based methods in the development of target tracking control scheme

# General control approaches



Differential Wheel Robot

• Kinematic Model

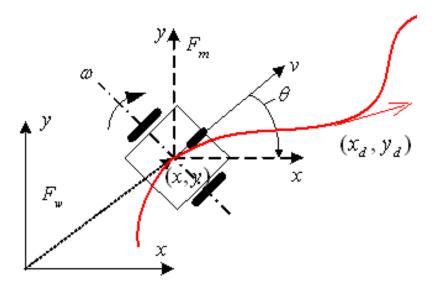
$$\begin{pmatrix} \dot{x} \\ \dot{y} \\ \dot{\theta} \end{pmatrix} = \begin{pmatrix} \cos \theta & 0 \\ \sin \theta & 0 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} v \\ \omega \end{pmatrix}$$

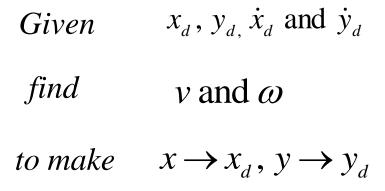
• *Nonholonomic Constraint* (rolling contact without slipping)

 $\dot{x}\sin\theta - \dot{y}\cos\theta = 0$ 

■Nonhonolonic (No-integrable) and under actuated (2-input~3-output) ■cannot be stabilized by time-invariant or smooth feedback control

#### Trajectory tracking (Cartesian coordinates based)

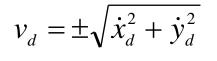




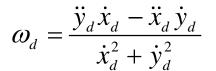
It can be proved (due to Lyapunov and Barbalat), the following control can meet the objective :

$$v = v_d \cos(\theta_d - \theta) + k_1 [\cos\theta(x_d - x) + \sin\theta(y_d - y)]$$

 $\omega = \omega_d + k_2 \operatorname{sgn}(v_d) [\sin \theta (x_d - x) - \cos \theta (y_d - y)] + k_3 (\theta_d - \theta)$ 



Desired linear velocity (along the trajectory)



Desired angular velocity

 $\theta_d = ATAN2(\dot{y}_d, \dot{x}_d) + k\pi$ 

Desired direction

$$k_1 = k_3 = 2\xi \sqrt{\omega_d^2 + bv_d^2}, \quad k_2 = b|v_d|$$

The trajectory needed to be specified in Note: prior; the controller fails when  $v_d = 0$ 

with nonlinear modifications to adjust angular motion:

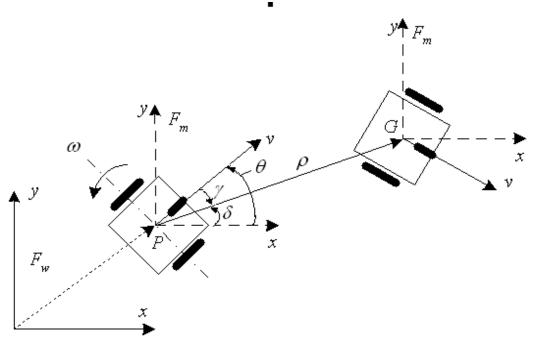
$$v = v_d \cos(\theta_d - \theta) + k_1 [\cos\theta(x_d - x) + \sin\theta(y_d - y)]$$

$$\omega = \omega_d + \bar{k}_2 v_d \frac{\sin(\theta_d - \theta)}{\theta_d - \theta} [\sin \theta (x_d - x) - \cos \theta (y_d - y)] + k_3 (v_d, \omega_d) (\theta_d - \theta)$$

where 
$$k_2 = b$$

,

#### Goal / target tracking (Polar coordinates based)



Control task: move the robot from its original posture:  $(x_p, y_p, \theta_p)$  to the target posture  $(x_g, y_g, \theta_g)$   $(\theta_g = 0: parallel parking).$ 

The system model described in polar coordinates:

$$\dot{\rho} = -v\cos\gamma, \ \dot{\gamma} = -\omega + v\frac{\sin\gamma}{\rho}, \ \dot{\delta} = v\frac{\sin\gamma}{\rho}$$
$$\rho = \sqrt{(x_g - x)^2 + (y_g - y)^2}$$
$$\delta = \tan^{-1}(\frac{y_g - y}{x_g - x})$$
$$\gamma = \delta - \theta$$

The model is singular at  $\rho = 0$ 

Let 
$$v = k_1 \rho \cos \gamma$$
  $\omega = k_2 a_2 \gamma + \frac{k_1 \sin 2\gamma}{2a_2 \gamma} (a_2 \gamma + a_3 \delta)$ 

It can be proved that ( *due to Lyapunov and Barbalat*)

$$\rho \rightarrow 0, \gamma \rightarrow 0, \delta \rightarrow 0$$

with the Lyapunov function candidate  

$$V = \frac{1}{2}a_1\rho^2 + \frac{1}{2}a_2\gamma^2 + \frac{1}{2}a_3\delta^2, a_1, a_2, a_3 > 0$$

$$\dot{V} = -a_1k_1\cos^2\gamma \ \rho^2 - k_2a_2^2\gamma^2 \le 0$$

• large control effort or fluctuation when the angle tracking error is near zero or the linear tracking error is big

• the target is assumed to be stationary

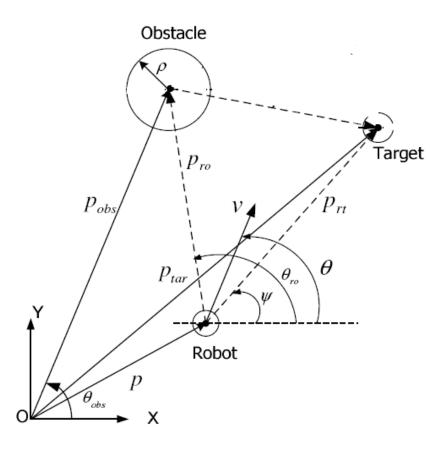
### Potential field approach (point mass model)

Attractive and repulsive fields:

$$U_{att} = \frac{1}{2} \xi_1 p_{rt}^T p_{rt}$$
$$U_{rep} = \begin{cases} \frac{1}{2} \xi_2 (\rho^{-1} - \rho_0^{-1})^2, & \text{if } \rho \le \rho_0 \\ 0 & \text{else} \end{cases}$$

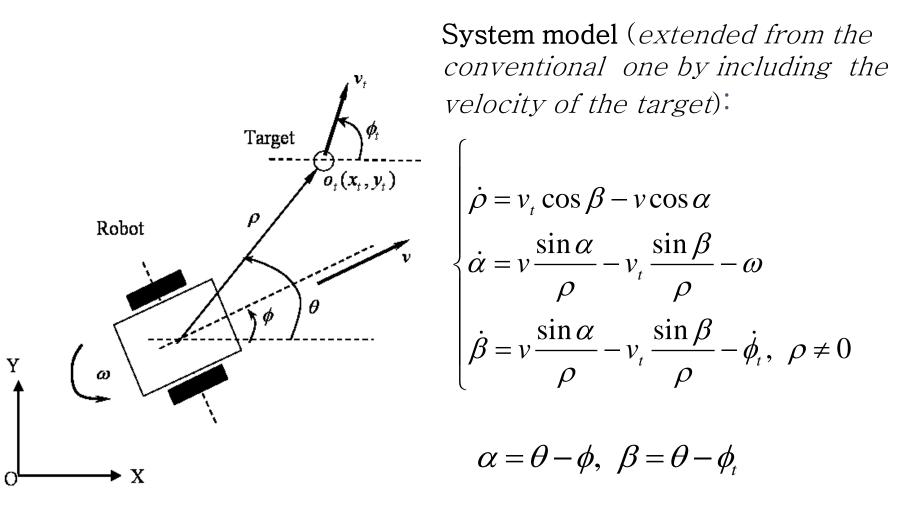
Robot move along the negative gradient of the combined field:

$$v = -\nabla_{p} U_{att}(p) - \nabla_{p} U_{rep}(p)$$
  
= 
$$\begin{cases} \xi_{1} p_{rt} + \xi_{2} (\rho^{-1} - \rho_{0}^{-1}) \rho^{-2} \nabla_{p} \rho, \text{ if } \rho \leq \rho_{0} \\ \xi_{1} p_{rt}, & \text{else} \end{cases}$$



- •The law only specifies the direction of the robot velocity
- target is assumed to be stationary
- local minima

## Lyapunov based target tracking controller with limited control efforts



#### Controller 1: Extension of the general control approach

$$v = (v_t \cos \beta + \lambda_v \rho) \cos \alpha$$
$$\omega = \lambda_\alpha \alpha + \frac{\alpha + \beta}{\rho} (\frac{\sin 2\alpha}{2\alpha} \cos \beta - \frac{\sin \beta}{\alpha}) v_t - \frac{\beta}{\alpha} \dot{\phi}_t + \frac{\sin 2\alpha}{2\alpha} \lambda_v (\alpha + \beta))$$

It can be proved with Lyapunov method, that under the controller,

 $\alpha \rightarrow 0, \rho \rightarrow 0 \text{ and } \beta \rightarrow 0$ 

(Lyapunov function candidate:  $V = \frac{1}{2}(\rho^2 + \alpha^2 + \beta^2)$ )

Note:

- target motions directly affects the control efforts
- sinusoidal functions of the systems states attenuate the magnitude of control
- tracking errors appear as the denominators in the terms of the controller
- linear tracking and angular tracking errors are treated equally too demanding ?

#### Controller 2: Improvement from Controller 1

Prioritise and change the control objectives:

 $\rho \rightarrow 0, \ \alpha \rightarrow 0 \text{ (or bounded)}, \ \alpha - \beta \rightarrow 0 \text{ (or bounded)}$ 

and reflect them in the definition of the Lyapunov function:

$$V = \frac{1}{2}\rho^{2} + \frac{1}{2}\alpha^{2} + \frac{1}{2}(\alpha - \beta)^{2}$$

New controller:

$$v = (v_t \cos \beta + \lambda_v \rho) \cos \alpha$$
$$\omega = \frac{\alpha}{2\alpha - \beta} (\lambda_\alpha \alpha + \frac{v_t}{\rho} (\frac{\sin 2\alpha}{2} \cos \beta - \sin \beta) + \frac{\sin 2\alpha}{2} \lambda_v) + \frac{\alpha - \beta}{2\alpha - \beta} \dot{\phi}_t$$

which can also achieve the convergence of the tracking errors, but with less control efforts

Comparison: control efforts of Controllers 1 and Controller 2

$$\omega_{1} = \lambda_{\alpha}\alpha + \frac{\alpha + \beta}{\rho}(\frac{\sin 2\alpha}{2\alpha}\cos\beta - \frac{\sin\beta}{\alpha})v_{t} - \frac{\beta}{\alpha}\dot{\phi}_{t} + \frac{\sin 2\alpha}{2\alpha}\lambda_{v}(\alpha + \beta)$$

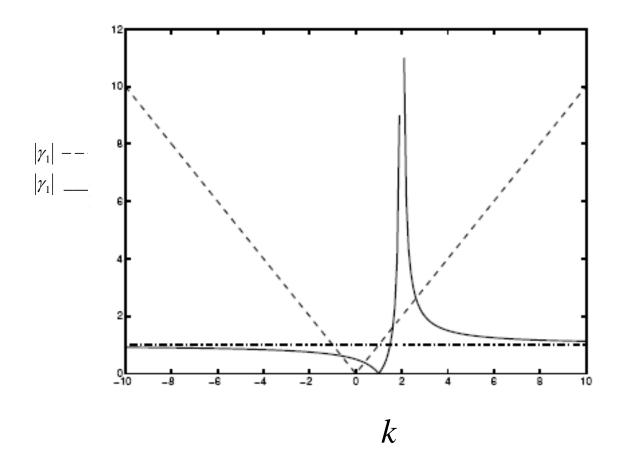
$$\omega_2 = \frac{\alpha}{2\alpha - \beta} (\lambda_\alpha \alpha + \frac{v_t}{\rho} (\frac{\sin 2\alpha}{2} \cos \beta - \sin \beta) + \frac{\sin 2\alpha}{2} \lambda_v) + \frac{\alpha - \beta}{2\alpha - \beta} \dot{\phi}_t$$

or

$$\omega_1 = \xi - \gamma_1 \eta - \lambda_\alpha \beta$$
$$\omega_2 = \xi - \gamma_2 \eta$$

$$\xi = \lambda_{\alpha} \alpha + \frac{v_t}{\rho} (\frac{\sin 2\alpha}{2} \cos \beta - \sin \beta) + \frac{\sin 2\alpha}{2} \lambda_v$$
$$\eta = \xi - \dot{\phi}_t, \ \gamma_1 = k, \ \gamma_2 = \frac{1-k}{2-k}$$
$$k = \frac{\beta}{\alpha}$$

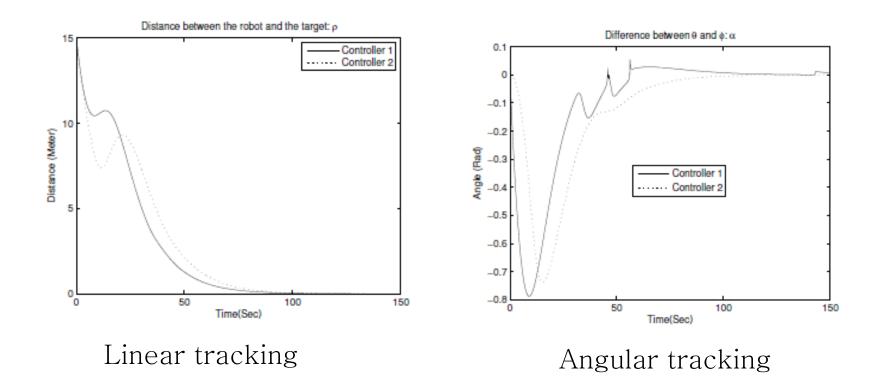
- By observation, the magnitude of controller 2 is less than that controller 1
- Analysing the factors ( $\gamma$ ) affecting the controller magnitude, it is obvious that, except for the region near  $k = \frac{\beta}{\alpha} = 2$  that affecting Controller 1 is larger in magnitude than that affecting Controller 2.

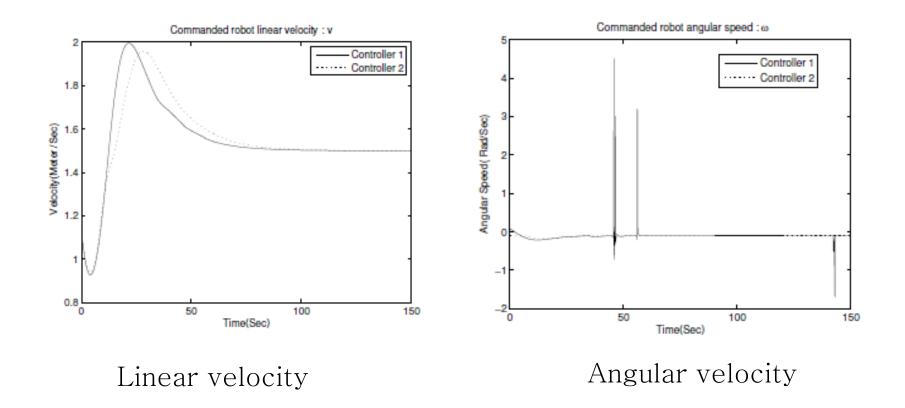


Seminar at Monash University, Sunway Campus, 14 Dec 2009

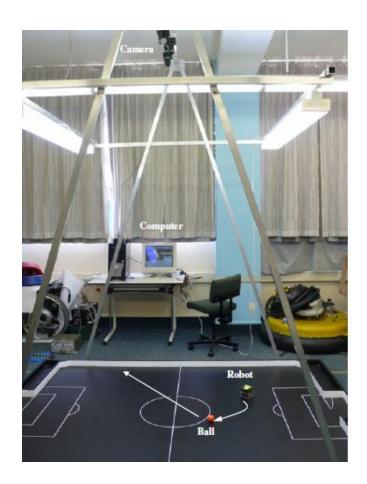
Simulation Results (tracking a target Moving along a circle)

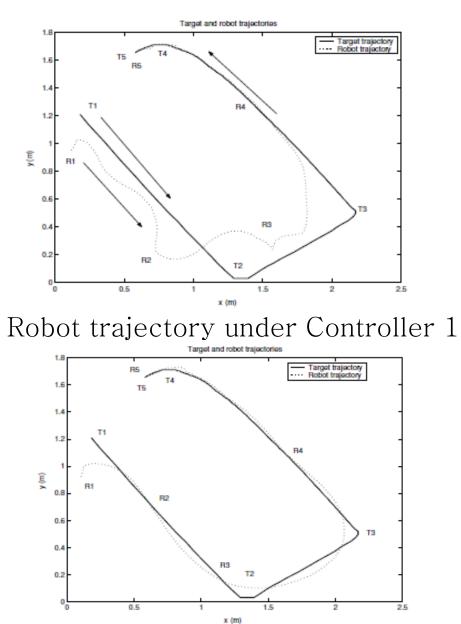
$$x_t = 3 - 15\cos(0.08t), \ y_t = 47 + 15\sin(0.08t), \ v_t = 1.2$$
  
 $\lambda_v = 0.075, \ \lambda_{\alpha} = 0.15$ 





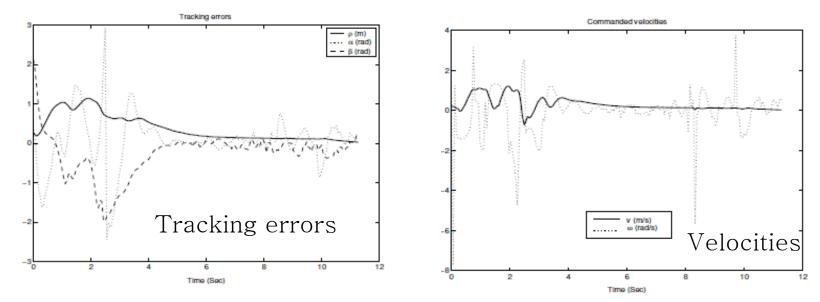
#### <u>Experiments</u>



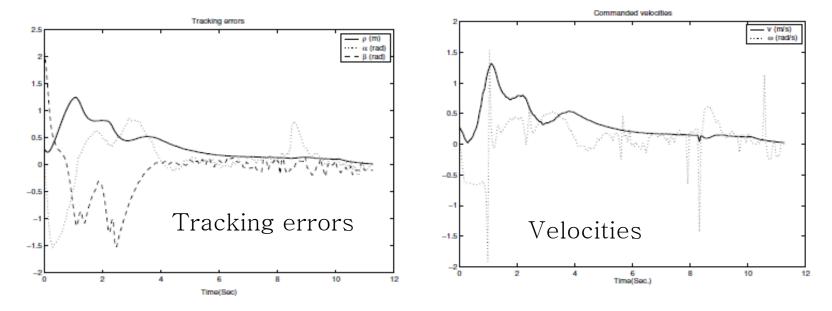


Robot trajectory under Controller 2

#### Under Controller 1:



Under Controller 2:



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#### Demonstrations





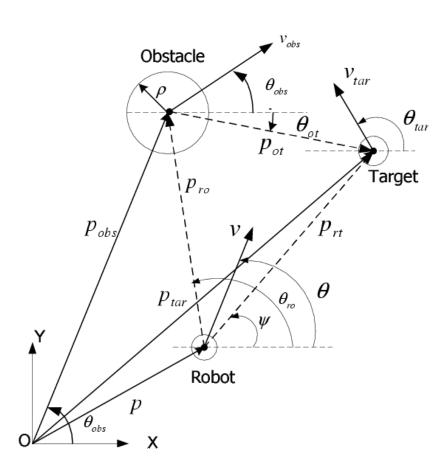
#### Controller 1

#### Controller 2

#### Conclusions:

- It is feasible to reduce the control efforts through
  - prioritization of control objectives
  - > defining of Lyapunov function to reflect that priority
  - attenuation of controller outputs with some special functions of the system states (like sinusoidal functions etc.)
     while achieving the same or better control results in comparison with the conventional controllers
- The performance of the controller is affected by the noises of the sensors for state feedback (esp. velocity).

# Potential field based control approach for robot's target tracking



System model:

$$p_{rt} = \begin{bmatrix} x_{rt} & y_{rt} \end{bmatrix}^{T}$$
$$\dot{x}_{rt} = v_{tar} \cos \theta_{tar} - v \cos \theta$$
$$\dot{y}_{tar} = v_{tar} \sin \theta_{tar} - v \sin \theta$$

Potential fields:

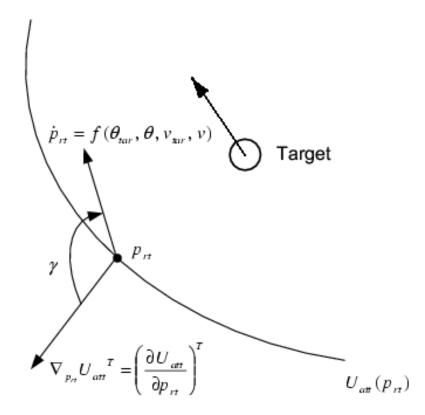
$$U = U_{att} + U_{rep}$$

$$U_{att} = \frac{1}{2} \xi_1 p_{rt}^T p_{rt}$$

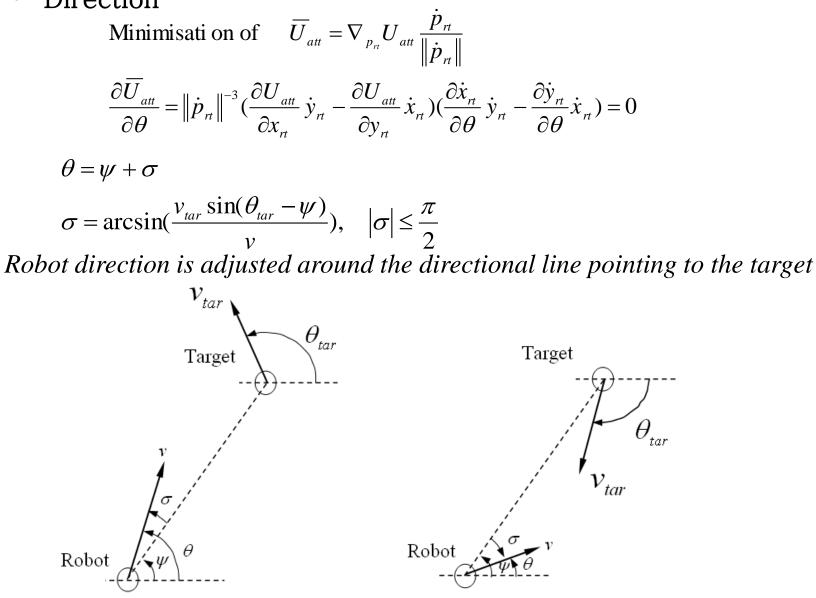
$$U_{rep} = \begin{cases} \frac{1}{2} \xi_2 (\rho^{-1} - \rho_0^{-1})^2, & \text{if } \rho \le \rho_0 \\ 0 & \text{else} \end{cases}$$

#### Case 1: Moving target free of obstacles

Minimization of the angle between the gradient of the field and the direction of robot motion relative to the target.



• Direction



The target moves away from the robot

The target moves to the robot

• Speed Intuitively  $v \ge \|v_{tar} \sin(\theta_{tar} - \psi)\|$ 

It is chosen to decrease 
$$U_{att}$$
, or  
 $\dot{U}_{att} = \xi_1 p_n^T \dot{p}_n = \xi_1 \|p_n\| (v_{tar} \cos(\theta_{tar} - \psi) - v \cos \sigma)$   
 $= \xi_1 \|p_n\| (v_{tar} \cos(\theta_{tar} - \psi) - (v^2 - v_{tar}^2 \sin^2(\theta_{tar} - \psi))^{\frac{1}{2}}) < 0$ 

One of the choices is:

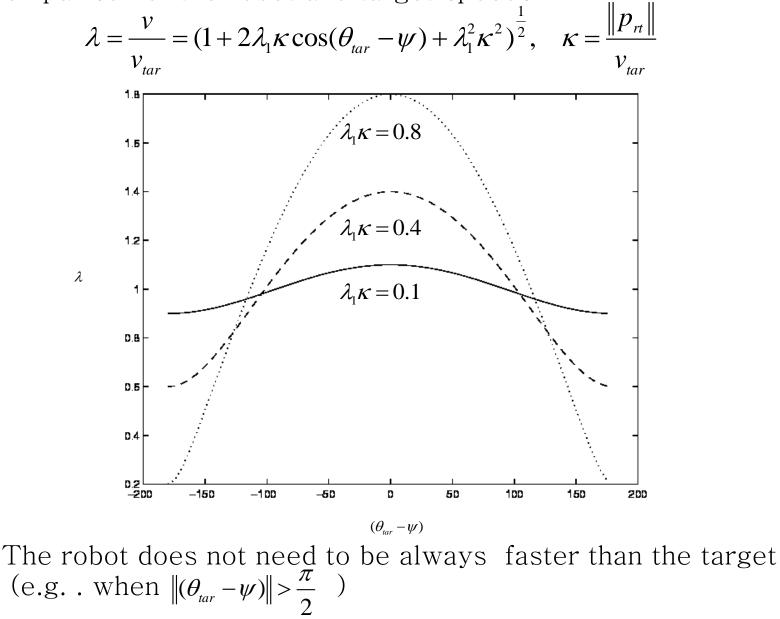
$$v = (v_{tar}^{2} + 2\lambda_{1}v_{tar} \|p_{rt}\|\cos(\theta_{tar} - \psi) + \lambda_{1}^{2} \|p_{rt}\|^{2})^{\frac{1}{2}}$$

It leads to:

$$U_{att} = U_{att}(0)e^{-2\lambda_1 t} \to 0$$
$$\|p_{rt}\| \to 0$$

The speed determined by the relative linear distance, the target velocity and there directional relationship.

Comparison of the robot and target speeds:



#### Case 2: Moving target with moving obstacles

The approach can be extended to solve the path/speed planning of the robot surrounded by multiple obstacles.

$$\theta = \overline{\psi} + \arcsin\frac{v_{tar}\sin(\theta_{tar} - \overline{\psi})}{v}$$

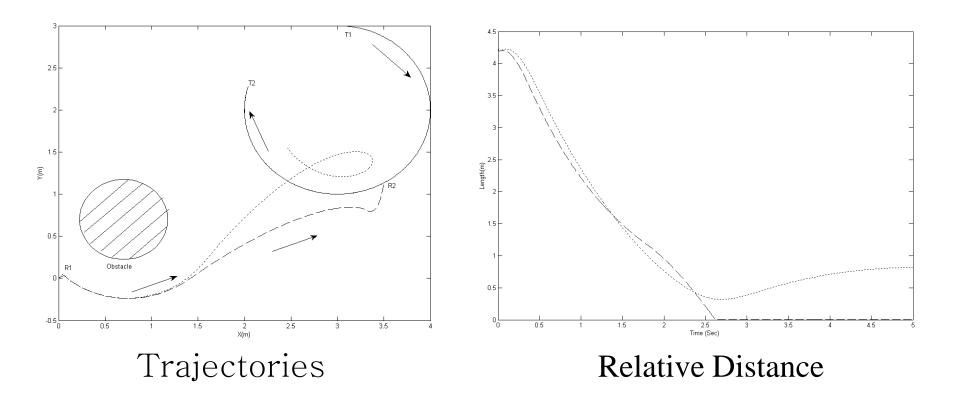
$$v = \sqrt{v_{tar}\cos(\theta_{tar} - \psi) - \sum_{i=1}^{n}\beta_{i}v_{obsi}\cos(\theta_{obsi} - \theta_{roi}) + \lambda_{1}\|p_{rt}\|)^{2} + v_{tar}^{2}\sin^{2}(\theta_{tar} - \overline{\psi})}$$

$$\overline{\psi} = \arctan\frac{\sin\psi - \sum_{i=1}^{n}\beta_{i}\sin\theta_{roi}}{\cos\psi - \sum_{i=1}^{n}\beta_{i}\cos\theta_{roi}}$$

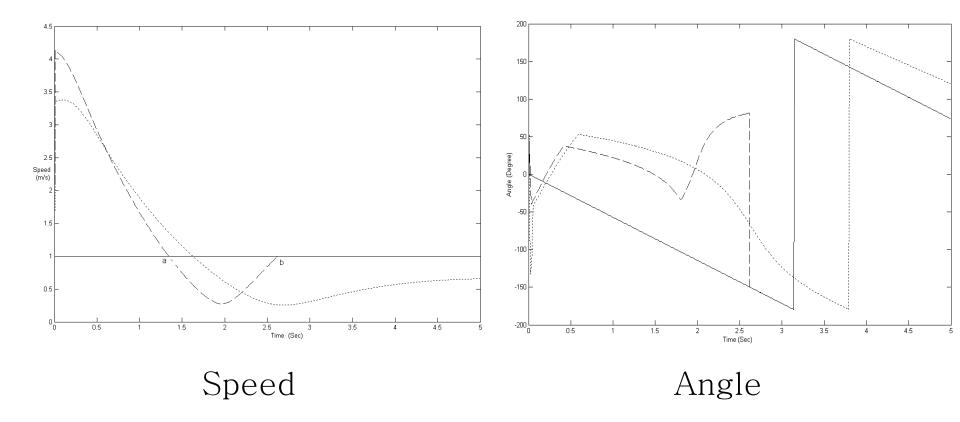
$$\beta_{i} = \frac{\eta_{i}\|p_{roi}\|}{\xi_{1}\|p_{rt}\|}, \quad \eta_{i} = \xi_{2}\rho_{i}^{-2}\|p_{roi}\|^{-1}(\rho_{i}^{-1} - \rho_{0}^{-1})$$

#### Simulation Results:

 $x_t = 3.0 + \sin t, y_t = 2.0 + \cos t$  $\theta_{tar} = -t, \quad v_{tar} = 1.0, \quad \lambda_1 = 1$ 

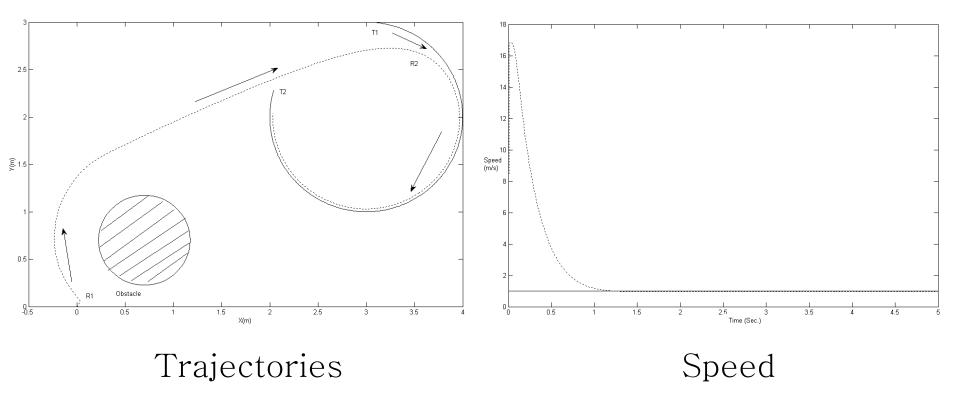


Solid line: target Dashed line : robot under the proposed controller Dotted line :robot under the conventional potential field controller



Solid line: target Dashed line : robot under the proposed controller Dotted line :robot under the conventional potential field controller

#### Performance of the conventional field method with a high gain

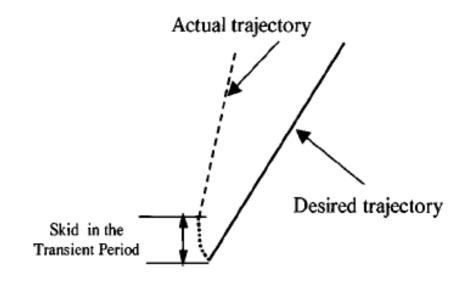


### Conclusion:

- the speed as well as the direction of the robot motion are determined with potential field method
- the velocity of the moving target is taken into consideration
- the proposed approach maintains or improves tracking accuracy and reduce control efforts, in comparison to the traditional approaches
- further study on the determination of the optimum speed of the robot can be done by specifying additional performance requirements.

## Speed control considering dynamic coupling between the actuators

- Synchronisation of the wheels' motion affects the robot's trajectory
- Coupling between the actuators needs to be considered



#### Model based adaptive control

Dynamic model:  $M\dot{\omega} + \beta C(\omega) = \tau$  $\omega = [\omega_r \quad \omega_r]^T$  $M = \begin{vmatrix} \frac{r^2}{4b^2}(mb^2 + I) + I_w & \frac{r^2}{4b^2}(mb^2 - I) \\ \frac{r^2}{4b^2}(mb^2 - I) & \frac{r^2}{4b^2}(mb^2 + I) + I_w \end{vmatrix}$  $C(\omega) = \begin{vmatrix} 0 & \omega_r - \omega_l \\ \omega_r - \omega_r & 0 \end{vmatrix} \qquad \beta = \frac{m_c dr^3}{4b^2}, \quad m = m_c + 2m_w$  $I = m_c d^2 + 2m_w b^2 + I_c + 2I_w$ 

 $m_c, m_w, I_m, I_c$  are the inertia parameters of the robot and the wheels b, d, r are the geometric parameters

Introducing new variables  $\omega' = [\omega'_1 \quad \omega'_2]^T, \ \tau' = [\tau'_1 \quad \tau'_2]^T$   $\omega'_1 = \omega_r + \omega_l, \ \omega'_2 = \omega_r - \omega_l$   $\tau'_1 = \tau'_r + \tau_l, \ \tau'_2 = \tau'_r - \tau_l$ then  $\omega = T\omega', \ \tau = \frac{1}{2} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$ 

Dynamic model is transformed to a more compact form :

 $M'\dot{\omega}' + \beta \omega_2' C' \omega' = \tau'$ 

$$M' = T^{-1}MT = \begin{bmatrix} \alpha_1 & 0 \\ 0 & \alpha_2 \end{bmatrix}, \quad C' = \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix}$$
$$mr^2 \qquad \qquad Ir^2$$

$$\alpha_1 = \frac{m}{2} + I_w, \ \alpha_2 = \frac{m}{2b^2} + I_w$$

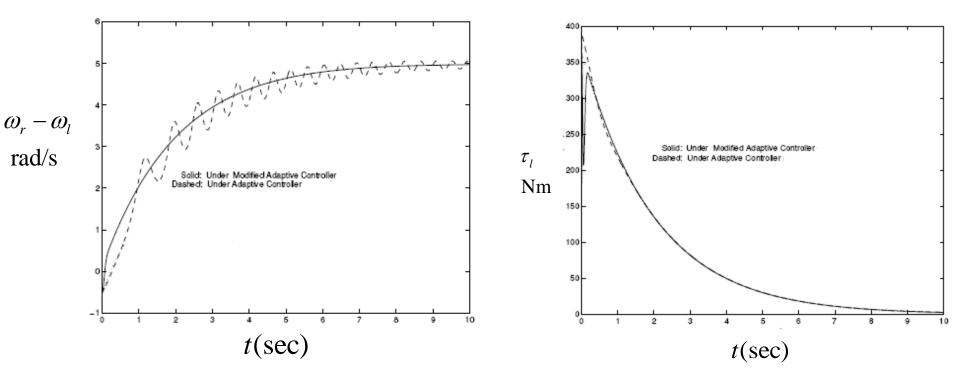
Based on the transformed dynamic model, the adaptive speed controllers are derived:

$$\begin{aligned} \tau_{r} &= \frac{k_{1} + k_{2}}{2} (\omega_{rd} - \omega_{r}) + \frac{k_{1} - k_{2}}{2} (\omega_{ld} - \omega_{l}) + \frac{1}{2} (\hat{\alpha}_{1} \dot{\omega}_{ld}' + \hat{\alpha}_{2} \dot{\omega}_{2d}') + \hat{\beta} \omega_{2}' \omega_{ld} \\ \tau_{l} &= \frac{k_{1} + k_{2}}{2} (\omega_{ld} - \omega_{l}) + \frac{k_{1} - k_{2}}{2} (\omega_{rd} - \omega_{r}) + \frac{1}{2} (\hat{\alpha}_{1} \dot{\omega}_{ld}' - \hat{\alpha}_{2} \dot{\omega}_{2d}') - \hat{\beta} \omega_{2}' \omega_{rd} \\ \dot{\hat{\alpha}}_{1} &= -\gamma \dot{\omega}_{1d} e_{1}, \ \dot{\hat{\alpha}}_{2} &= -\gamma \dot{\omega}_{2d} e_{2}, \ \dot{\hat{\beta}} &= -\gamma \omega_{2}' (\omega_{2}' \omega_{1d} - \omega_{1}' \omega_{2d}) \\ e_{1} &= \omega_{1d}' - \omega_{1}, \ e_{2} &= \omega_{2d}' - \omega_{2} \end{aligned}$$

Modified to reduce the amplitudes of the control outputs:

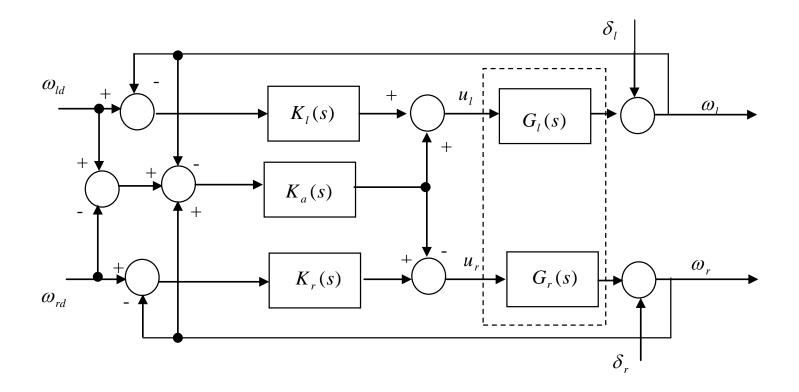
$$\begin{split} \tau_{r} &= \frac{k_{1} + k_{2}}{2} (\omega_{rd} - \omega_{r}) + \frac{k_{1} - k_{2}}{2} (\omega_{ld} - \omega_{l}) + \frac{1}{2} (\hat{\alpha}_{1} \dot{\omega}_{ld}' + \hat{\alpha}_{2} \dot{\omega}_{2d}') + k \hat{\beta} \omega_{l} \\ \tau_{l} &= \frac{k_{1} + k_{2}}{2} (\omega_{ld} - \omega_{l}) + \frac{k_{1} - k_{2}}{2} (\omega_{rd} - \omega_{r}) + \frac{1}{2} (\hat{\alpha}_{1} \dot{\omega}_{ld}' - \hat{\alpha}_{2} \dot{\omega}_{2d}') - k \hat{\beta} \omega_{r} \\ \dot{\hat{\beta}} &= -k \gamma (\omega_{2}' e_{1} - \omega_{1}' e_{2}) \\ k &= \omega_{2}' - \gamma_{k} (\omega_{2}' e_{1} - \omega_{1}' e_{2}) \end{split}$$

#### Simulation results



#### **Model free PID control**

A loop for the coupling of the wheels' speeds is added.



When  $G_{l}(s) = G_{r}(s) = G(s), K_{l}(s) = K_{r}(s) = K(s)$ 

Transfer functions :

$$\begin{split} G_{ind}(S) &= \frac{G(s)K(s)(K(s)+2K_s(s))}{G(s)K(s)(K(s)+2K_a(s))+K(s)+K_a(s)},\\ G_{ind}(S) &= \frac{K_s(s)}{G(s)K(s)(K(s)+2K_a(s))+K(s)+K_a(s)}, \end{split}$$

$$\omega_{l}(s) = G_{ind}(s)\omega_{ld}(s) - G_{syn}(s)\omega_{r}(s)$$
$$\omega_{r}(s) = G_{ind}(s)\omega_{rd}(s) - G_{syn}(s)\omega_{l}(s)$$

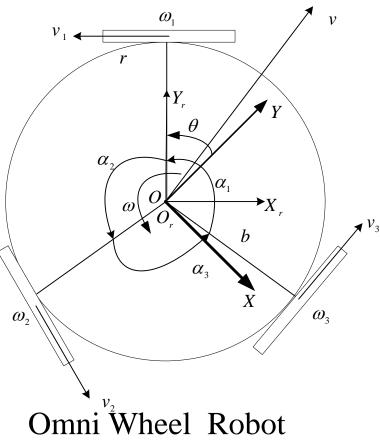
• First order motor model is adopted:

$$G(s) = \frac{k_m}{1 + \tau_m s}, \ \tau_m = JR_a / K_t^2$$

- PID controller is used for the speed control
- Implemented with one PIC18F252 microcontroller

#### Speed Control of an Omni-wheel robots

Modeling (Kinematics)



Inverse kinematic model:

$$r\omega_i = b\omega + v_r v_i \quad (i = 1, 2, 3)$$

$$v_1 = [-1 \ 0]^T$$

$$v_2 = [\cos\frac{\pi}{3} - \sin\frac{\pi}{3}]^T$$

$$v_3 = [\cos\frac{\pi}{3} - \sin\frac{\pi}{3}]^T$$

$$\omega_{1} = r^{-1}(b\omega - v_{rx})$$

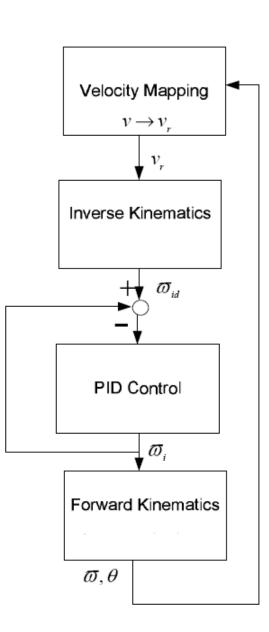
$$\omega_{2} = r^{-1}(b\omega + v_{rx}\cos\frac{\pi}{3} - v_{ry}\sin\frac{\pi}{3})$$

$$\omega_{3} = r^{-1}(b\omega + v_{rx}\cos\frac{\pi}{3} + v_{ry}\sin\frac{\pi}{3})$$

$$v_{r} = [v_{rx} \quad v_{ry}]^{T}$$

$$v_{rx} = v_{x}\cos\theta + v_{y}\sin\theta$$

$$v_{ry} = -v_{x}\sin\theta + v_{y}\cos\theta$$

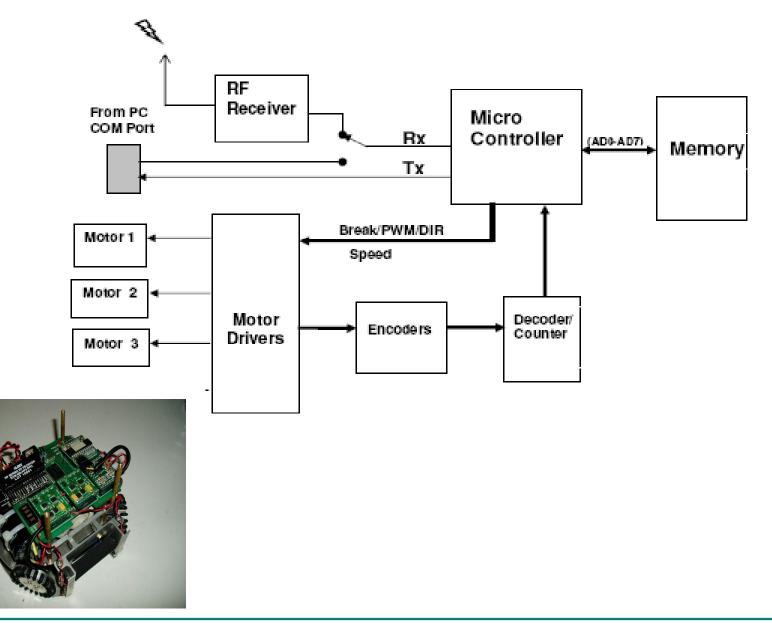


- Chooped fed motors with drivers to drive the wheels
- PID controller implemented with one one 80296 microcontrollers (three PWM outputs)
- Encoder resolution 512 ppr
- Sampling time 1 ms
- Control loop completed within 0.5ms

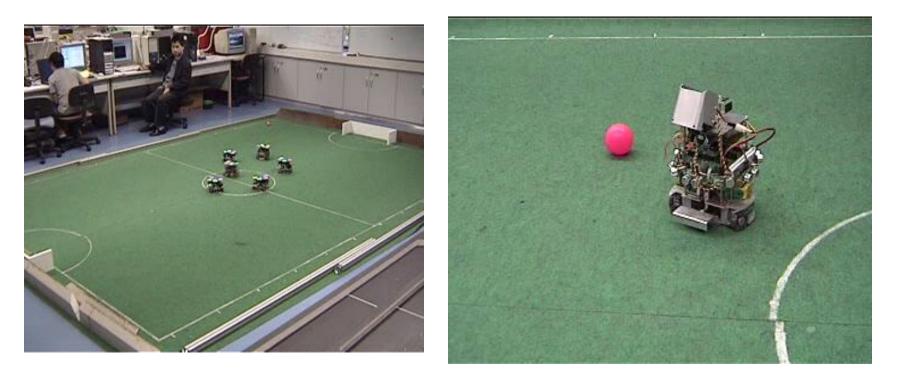
This is achieved through:

- codes written in an assembly language without using floating point libraries (too slow)
- fixed point notation and a look up table of whole numbers to represent a floating point number with reasonable accuracy
- only the simple operations like addition, substration, multiplication and bits-shifting are used.

## Implementation



# Demonstrations



## Conclusion

- Lyapunov and potential field based target tracking controllers, and speed controller for dynamically coupled wheels for mobile robots were presented
- Both position and velocity of the target were considered in the target tracking controller design
- Functions of the system states, especially those of the target, are are designed to moderate the magnitude or fluctuation of the control effort
- The states of the system were assumed to be available; sensor noises affect the performance of the controller.
- To get a good system states estimation and prediction from the sensor data is another big issue to be addressed together with the controller design (*Kalman filtering, Bayesian method* etc.)
- Further study can be undertaken on integrating open-loop optimal control, closed-loop control and system states estimation and prediction